Twice -Band Microstrip Hexagonal Slotted Patch Antenna for Microwave Communication

Bipa Datta^{#1}, Arnab Das ^{#2}, Samiran Chatterjee^{#3},

^{#1,2,3}ECE Deptt., Brainware Group of Institutions, Barasat, West Bengal University of Technology, Kolkata, West

Bengal, India

¹bipa.datta@gmail.com, ²u_call_arnab@yahoo.co.in, ³samiranengineer@gmail.com

Moumita Mukherjee^{*4},

^{*4}Centre for Millimeter wave Semiconductor Devices and Systems, University of Calcutta, West Bengal, India ⁴mm_drdo@yahoo.com

Santosh Kumar Chowdhury^{^5}

⁵West Bengal University of Technology, JIS College of Engineering, West Bengal, India
⁵santoshkumarchowdhury@gmail.com

Abstract—A single layer, single feed compact slotted patch antenna is thoroughly simulated in this paper. Resonant frequency has been reduced drastically by cutting three equal slots which are same hexagonal structure at the upper right, upper left and lower left corner from the conventional microstrip patch antenna. It is shown that the simulated results are in acceptable agreement. More importantly, it is also shown that the differentially-driven microstrip antenna has higher gain of simulated 3.24 dBi at 9.91 GHz and 0.14 dBi at 13.61 GHz and beamwidth of simulated 163.19⁰ at 9.91 GHz and 122.10⁰ at 13.61 GHz of the single-ended microstrip antenna. Simulated antenna size has been reduced by 48.11% with an increased frequency ratio when compared to a Conventional microstrip patch antenna.

_ _ _ _ _ _ _ _ _ _ _ _

Index Terms — Compact, Patch, Slot, Resonant frequency, Bandwidth.

_ _ _ _ _ _ _ _ _ _ _ _

1. INTRODUCTION

N recent years, demand for small antennas on wireless L communication has increased the interest of research work on compact microstrip antenna design among microwave and wireless engineers [1-6]. Microstrip antennas have many unique and attractive properties - low in profile, light in weight, compact and conformable in structure, and easy to fabricate to support the high mobility necessity for a wireless telecommunication device and for high resolution mapping for radar communication, a small and light weight compact microstrip antenna is one of the most suitable application. The development of antenna for wireless communication also requires an antenna with more than one operating frequency. This is due to many reasons, primarily because of various wireless communication systems and many telecommunication operators use various frequencies. In its most fundamental form, a Microstrip Patch antenna consists of a radiating patch on one side of a dielectric substrate which has a ground plane on the other side. Therefore one antenna that has multiband characteristic is more desirable than having one antenna for each frequency band.

Most effective technique is cutting slot in proper position on the microstrip patch. In this paper includes cutting three equal slots which are same hexagonal structure at the upper right, upper left and lower left corner from the conventional microstrip patch antenna, to increase the return loss and gainbandwidth performance of the simulated antenna (Figure 2). The antenna has become a necessity for many applications in recent wireless communication such as radar, microwave and space communication. To reduce the size of the antenna substrates are chosen with higher value of dielectric constant [7-9]. Our aim is to reduce the size of the antenna as well as increase the operating bandwidth. The proposed antenna (substrate with ε_r = 4.4) has a gain of 3.24 dBi and presents a size reduction of 48.11% when compared to a conventional microstrip patch (10mm X 6mm). The simulation has been carried out by IE3D [14] software which uses the MoM method. Due to the small size, low cost and low weight this antenna is a good entrant for the application of X-Band microwave communication and Ku-Band RADAR communication.

The X band belongs to in the microwave radio region of the electromagnetic spectrum. defined It is bv an IEEE standard for radio waves and radar engineering with frequencies that ranges from 8.0 to 12.0 GHz. The X band is used for short range tracking, missile guidance, marine, radar and air bone intercept. Especially, it is used for radar communication ranges roughly from 8.29 GHz to 11.4 GHz. The Ku-Band belongs to in the microwave radio region of the electromagnetic is defined spectrum. It by an IEEE standard for radio waves and radar engineering with frequencies that ranges from 12.0 to 18.0 GHz [10-11]. The Ku

band [12] is used for high resolution mapping and satellite altimetry. Specially, Ku Band [13] is used for tracking the satellite within the ranges roughly from 12.87 GHz to 14.43 GHz.

2. ANTENNA DESIGN

The configuration of the conventional printed antenna is shown in Figure 1 with L=6 mm, W=10 mm, substrate (PTFE) thickness h = 1.6 mm, dielectric constant ε_r = 4.4. Coaxial probe-feed (radius=0.5mm) is located at W/2 and L/3.

Assuming practical patch width W= 10 mm for efficient radiation and using the equation [6],

$$f_r = \frac{c}{2W} \times \sqrt{\frac{2}{(1+\varepsilon_r)}} \qquad \dots 1$$

Where, c = velocity of light in free space. Using the following equation [9] we have determined the practical length L = 6 mm.

$$L=L_{eff} - 2\Delta L$$
 ...

where,

$$\frac{\Delta L}{h} = 0.412 \times \left[\frac{(\mathcal{E}_{reff} + 0.3) \times (W/h + 0.264)}{(\mathcal{E}_{reff} - 0.258) \times (W/h + 0.8)}\right] \qquad \dots 3$$
$$\mathcal{E}_{reff} = \left(\frac{\mathcal{E}_{r} + 1}{2}\right) + \frac{\mathcal{E}_{r} - 1}{\left(2 \times \sqrt{\left(1 + 12 \times \frac{h}{W}\right)}\right)} \qquad \dots 4$$
and
$$L_{eff} = \left[\frac{c}{2 \times f_{r} \times \sqrt{\mathcal{E}_{reff}}}\right] \qquad \dots 5$$

2

Where, L_{eff} = Effective length of the patch, $\Delta L/h$ =Normalized extension of the patch length, ϵ_{reff} = Effective dielectric constant.

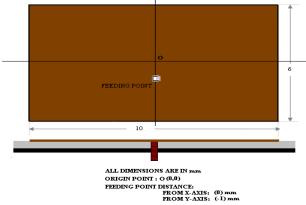




Figure 2 shows the configuration of simulated printed antenna designed with similar PTFE substrate. Three equal slots which are same hexagonal structure at the upper right, upper left

and lower left corner from the conventional microstrip patch antenna and the location of coaxial probe-feed (radius=0.5 mm) are shown in the figure 2.

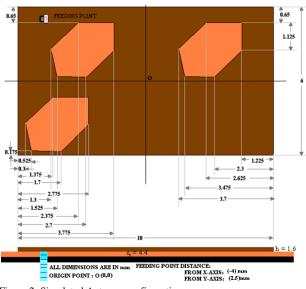
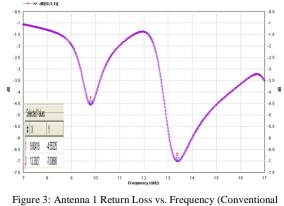


Figure 2: Simulated Antenna configuration

3. RESULTS AND DISCUSSION

Simulated (using IE3D [14]) results of return loss in conventional and simulated antenna structures are shown in Figure 3-4. A significant improvement of frequency reduction is achieved with simulated antenna compared to its conventional antenna counterpart.



Antenna)

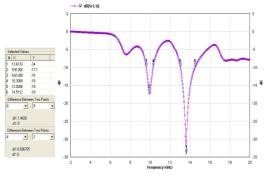


Figure 4: Slotted Antenna Return Loss vs. Frequency (Slotted Antenna)

In conventional antenna, return loss of about -7.0 dB is obtained at 13.39 GHz. Comparative analysis of Fig.3 &4 depicts that for the conventional antenna (fig.3), there is practically no resonant frequency at around 9.25 GHz with a return loss of around -6 dB. For the simulated antenna there is a resonant frequency at around 9.91 GHz with the return loss as high as -17.1 dB.

Due to the presence of slots in simulated antenna resonant frequency operation is obtained with large values of frequency ratio. The first and second resonant frequency is obtained at f_1 = 9.91 GHz with return loss of about -17.1 dB and at f_2 = 13.61 GHz with return losses -34dB respectively. Corresponding 10 dB bandwidth is obtained for Antenna 2 at f_1 and f_2 are 696.80 MHz and 1.48 GHz, respectively.

The simulated E plane and H-plane radiation patterns are shown in Figure 5-12. The simulated E plane (Total) radiation pattern of simulated antenna for 9.91 GHz is shown in figure 5.

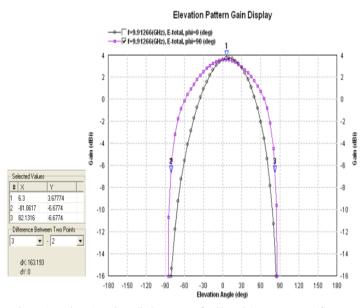


Figure 5: E-Plane (Total) Radiation Pattern for Slotted Antenna at 9.91 GHz

The simulated E plane (Total) radiation pattern (3D) of simulated antenna for 9.91 GHz is shown in figure 6.

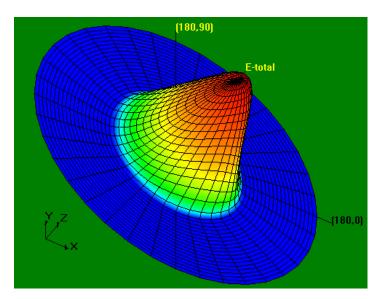


Figure 6: E-Plane (Total) Radiation (3D)Pattern for Slotted Antenna at 9.91 GHz

The simulated current distribution pattern of simulated antenna for 9.91 GHz is shown in figure 7.

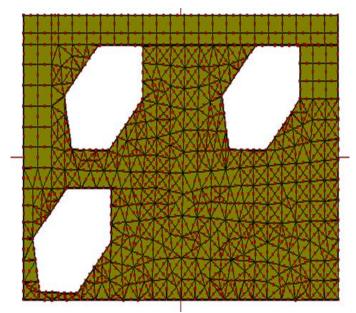


Figure 7: Current Distribution Pattern for Slotted Antenna at 9.91 GHz

The simulated E plane radiation pattern of simulated antenna for 9.91 GHz is shown in figure 8.

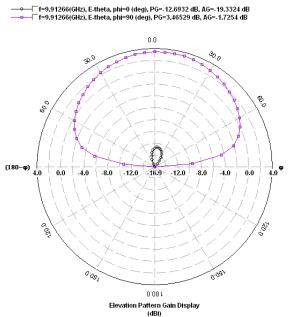


Figure 8: E-Plane Radiation Pattern for Slotted Antenna at 9.91 GHz

The simulated H plane radiation pattern of simulated antenna for 9.91 GHz is shown in figure 9.

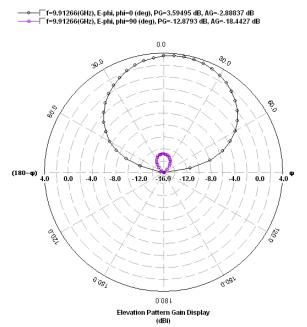


Figure 9: H-Plane Radiation Pattern for slotted Antenna at 9.91 GHz

The simulated E plane radiation pattern (3D-view) of Slotted Antenna for 9.91 GHz is shown in figure 10.

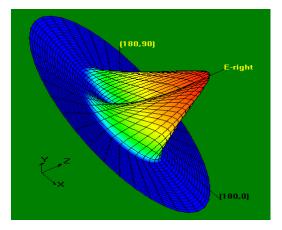


Figure 10: E-Plane Radiation Pattern (3D) for slotted antenna at 9.91 GHz

The simulated H plane radiation pattern (3D-view) of slotted antenna for 9.91 GHz is shown in figure 11.

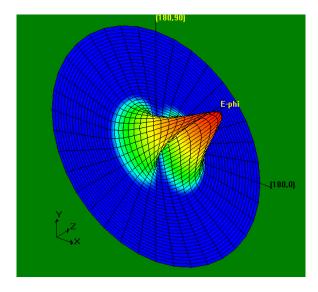


Figure 11: H-Plane Radiation Pattern (3D) for slotted antenna at 9.91 GHz

The simulated E plane (Total) radiation pattern of simulated antenna for 13.61 GHz is shown in figure 12.

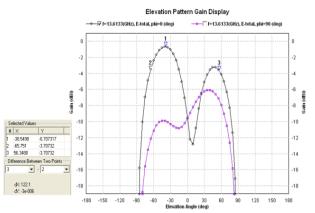


Figure 12: E-Plane (Total) Radiation Pattern for slotted antenna at 13.61GHz The simulated E plane (Total) radiation pattern (3D) of simulated antenna for 13.61 GHz is shown in figure 13.

IJSER © 2012 http://www.ijser.org

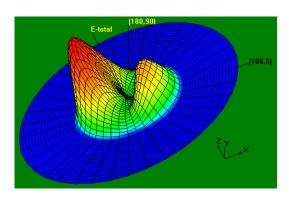


Figure 13: E-Plane (Total) Radiation Pattern (3D) for slotted antenna at 13.61 GHz

The simulated current distribution pattern of simulated antenna for 13.61 GHz is shown in figure 14.

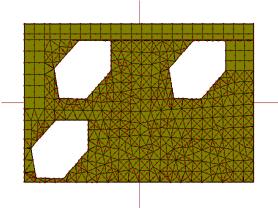


Figure 14: Current Distribution for slotted antenna at 13.61 GHz

The simulated E plane radiation pattern of simulated antenna for 13.61 GHz is shown in figure 15.

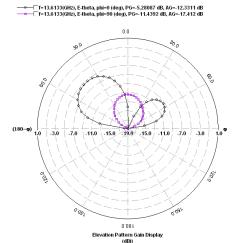


Figure 15: E-Plane Radiation Pattern for slotted antenna at 13.61GHz The simulated H plane radiation pattern of slotted antenna for 13.61 GHz is shown in figure 16.

→ T=13.6133(GHz), E-phi, phi=0 (deg), PG=-2.5373 dB, AG=-9.13185 dB → T=13.6133(GHz), E-phi, phi=90 (deg), PG=-6.88201 dB, AG=-14.1428 dB 0.0 0.0 0.0

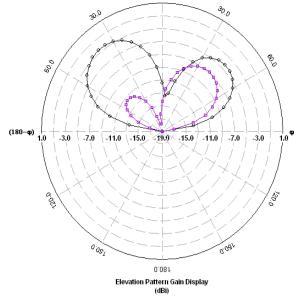


Figure 16: H-Plane Radiation Pattern for slotted antenna at 13.61GHz

The simulated E plane radiation pattern of slotted antenna (3D-view) for 13.61 GHz is shown in figure 17.

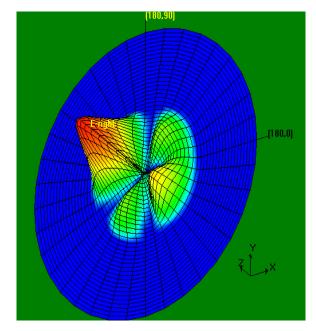


Figure 17: E-Plane Radiation Pattern (3D) for slotted antenna at 13.61 GHz

The simulated H plane radiation pattern of slotted antenna (3D-view) for 13.61 GHz is shown in figure 18.

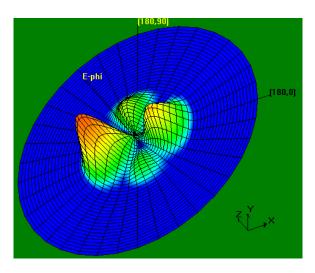


Figure 18: H-Plane Radiation Pattern (3D) for slotted antenna at 13.61 GHz

The simulated frequency vs. real part of the function for slotted antenna is shown in figure 19.

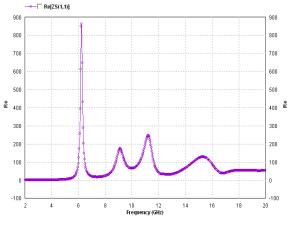
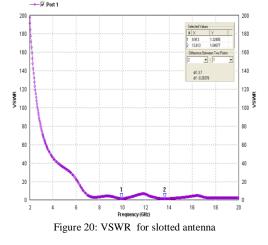


Figure 19: Frequency vs. real function for slotted antenna

The simulated frequency vs. VSWR for slotted antenna is shown in figure 20.



The simulated Smith Chart for slotted antenna is shown in figure 21.

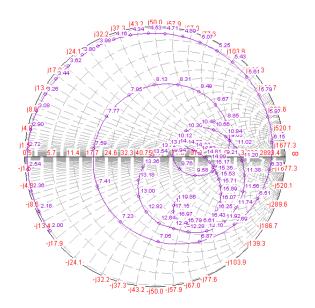


Figure 21: Smith Chart for slotted antenna

All the simulated results are summarized in the following Table1 and Table2.

TABLE I: Simulated results for antenna 1 and 2

ANTENNA STRUCTURE	RESONANT FREQUENCY (GHz)	RETURN LOSS (dB)	10 DB BAND- WIDTH (GHz)
Conventional	f1= 13.39	-7.00	NA
Slotted	f1= 9.91	-17.1	0.6968
	f ₂ = 13.61	-34	1.4826

TABLE II: Simulated results for antenna 1 and 2

ANTENNA STRUCTURE	RESONANT FREQUENCY (GHz)	3 DB BEAM- WIDTH (⁰)	ABSOLUTE GAIN (dBi)
Conventional	f1= 13.39	NA	NA
Slotted	f1= 9.91	163.190	3.24
	f ₂ = 13.61	122.1°	0.14
Frequency Ra	f ₂ / f ₁ = 1.373		

4. CONCLUSION

This paper focused on the simulated design on differentiallydriven microstrip antennas. Simulation studies of a single layer single feed micro strip printed antenna have been carried out using Method of Moment based software IE3D. Introducing slots at the edge of the patch size reduction of about 48.11% has been achieved. The 3dB beam-width of the radiation patterns are 163.19° (for f₁), 122.10° (for f₂) which is sufficiently broad beam for the applications for which it is intended.

The resonant frequency of slotted antenna, presented in the paper, designed for a particular location of feed point (-4mm, 2.5mm) considering the centre as the origin. Alteration of the location of the feed point results in narrower 10dB bandwidth and less sharp resonances.

ACKNOWLEDGEMENT(S)

S. K. Chowdhury acknowledges gratefully the financial support for this work provided by AICTE (India) in the form of a project entitled "DEVELOPMENT OF COMPACT, BROADBAND AND EFFICIENT PATCH ANTENNAS FOR MOBILE COMMUNICATION". M. Mukherjee wishes to acknowledge Defence Research and Development Organization (DRDO, Ministry of Defence), Govt. of India for their financial assistance.

REFERENCES

- [1] I.Sarkar, P.P.Sarkar, S.K.Chowdhury "A New Compact Printed Antenna for Mobile Communication", 2009 *Loughborough Antennas* & *Propagation Conference*, 16-17 November 2009, pp 109-112.
- [2] S. Chatterjee, U. Chakraborty, I.Sarkar, S. K. Chowdhury, and P.P.Sarkar, "A Compact Microstrip Antenna for Mobile Communication", IEEE annual conference. Paper ID: 510
- [3] J.-W. Wu, H.-M. Hsiao, J.-H. Lu and S.-H. Chang, "Dual broadband design of rectangular slot antenna for 2.4 and 5 GHz wireless communication", *IEE Electron. Lett.* Vol. 40 No. 23, 11th November 2004.
- [4] U. Chakraborty, S. Chatterjee, S. K. Chowdhury, and P. P. Sarkar, "A comact microstrip patch antenna for wireless communication," *Progress In Electromagnetics Research C*, Vol. 18, 211-220, 2011 <u>http://www.jpier.org/pierc/pier.php?paper=10101205</u>
- [5] Rohit K. Raj, Monoj Joseph, C.K. Anandan, K. Vasudevan, P. Mohanan, " A New Compact Microstrip-Fed Dual-Band Coplaner Antenna for WLAN Applications", *IEEE Trans. Antennas Propag.*, Vol. 54, No. 12, December 2006, pp 3755-3762.
- [6] Zhijun Zhang, Magdy F. Iskander, Jean-Christophe Langer, and Jim Mathews, "Dual-Band WLAN Dipole Antenna Using an Internal Matching Circuit", *IEEE Trans. Antennas and Propag.*, VOL. 53, NO. 5, May 2005, pp 1813-1818.
- [7] J. -Y. Jan and L. -C. Tseng, "Small planar monopole Antenna with a shorted parasitic inverted-L wire for Wireless communications in the 2.4, 5.2 and 5.8 GHz. bands", *IEEE Trans. Antennas and Propag.*, VOL. 52, NO. 7, July 2004, pp -1903-1905.
- [8] Samiran Chatterjee, Joydeep Paul, Kalyanbrata Ghosh, P. P. Sarkar and S. K. Chowdhury "A Printed Patch Antenna for Mobile Communication", Convergence of Optics and Electronics conference, 2011, Paper ID: 15, pp 102-107
- [9] C. A. Balanis, "Advanced Engineering Electromagnetics", John Wiley & Sons., New York, 1989.

- [10] Bipa Datta, Arnab Das, Samiran Chatterjee, Moumita Mukherjee, Santosh Kumar Chowdhury, " A Printed Microstrip Antenna for RADAR Communication," IOSR Journal of Electronics and Communication Engineering (IOSR-JECE), ISSN: (2278-2834), ISBN: (2278-8735), Vol. 3, Issue-5, pp 01-04, (Sep-Oct. 2012).
- [11] Bipa Datta, Arnab Das, Samiran Chatterjee, Bipadtaran Sinhamahapatra, Supriya Jana, Moumita Mukherjee, Santosh Kumar Chowdhury, "Design of Compact Patch Antenna for Multi-Band Microwave Communication", National Conference on Sustainable Development through Innovative Research in Science and Technology (Extended Abstracts), Paper ID: 115, pp 155, 2012
- [12] Arnab Das, Bipa Datta, Samiran Chatterjee, Bipadtaran Sinhamahapatra, Supriya Jana, Moumita Mukherjee, Santosh Kumar Chowdhury, "Multi-Band Microstrip Slotted Patch Antenna for Application in Microwave Communication," *International Journal of Science and Advanced Technology*, (ISSN 2221-8386), Vol. 2, Issue-9, 91-95, September 2012
- [13] Arnab Das, Bipa Datta, Samiran Chatterjee, Moumita Mukherjee, Santosh Kumar Chowdhury, "Multi-resonant Slotted Microstrip Antenna for C, X and Ku-Band Applications," IOSR Journal of Electrical and Electronics Engineering (IOSR-JEEE), ISSN: (2278-1676), Vol. 2, Issue-6, pp 47-52, (Sep-Oct. 2012).
- [14] Zeland Software Inc. IE3D: MoM-Based EM Simulator. Web: http://www.zeland.com/